

Digital Low-level RF Test at Diamond Booster

LLRF9 demo, March 16–20, 2015

D. Teytelman², et. al.

²Dimtel, Inc., San Jose, CA, USA

April 15, 2015

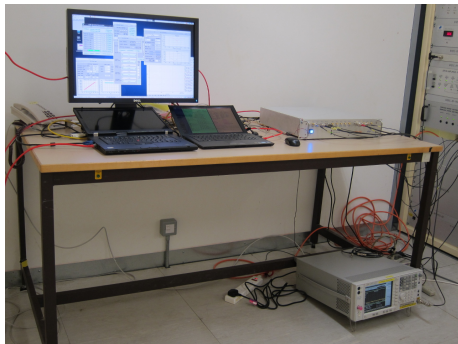


Outline

- 1 Setup
 - LLRF9 Introduction
 - Demo Setup and Schedule
- 2 System Operation
- 3 Performance Tests



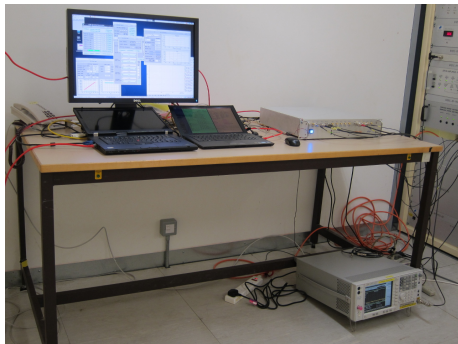
LLRF9 System



- A single 2U chassis for one- and two-cavity RF control;
- 9 input RF channels, 2 RF outputs;
- Tuner motor control via RS-485/Ethernet/EPICS;
- External interlock daisy-chain;
- Two external trigger inputs;
- Eight opto-isolated baseband ADC channels for slow interlocks.



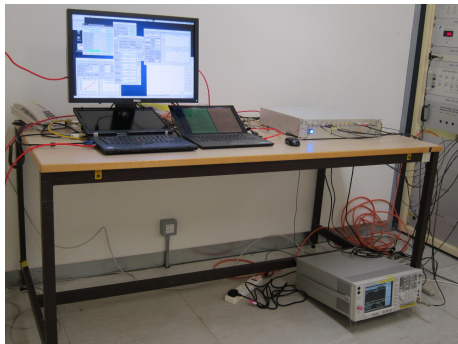
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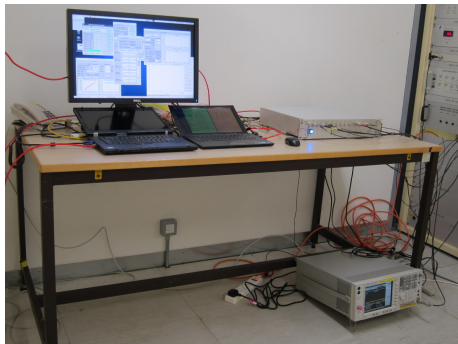
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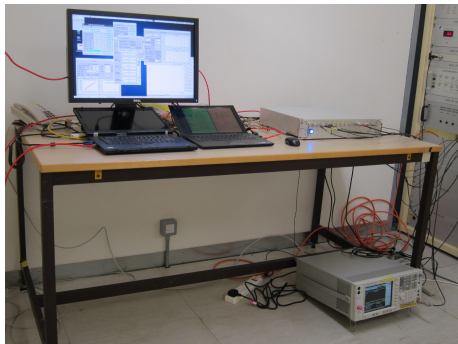
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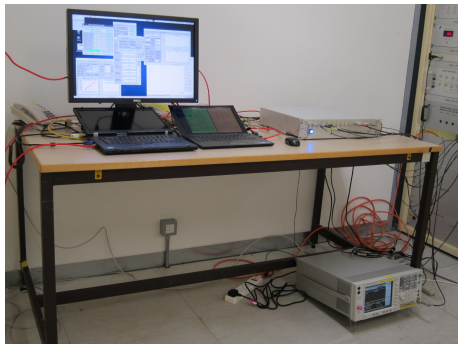
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Demo Setup

- Set up LLRF9 to run the booster RF with the following signals:
 - RF reference (500 MHz)
 - Three cavity probe signals (500 MHz)
 - Cavity forward signal (500 MHz)
 - Cavity reflected signal (500 MHz)
 - Drive output (500 MHz)
 - Interlock input (TTL)
 - Ramp trigger (TTL)
- Used EPICS interface to control cavity tuning;
- LLRF9 drive was connected to RF Safety Crate (113S) to maintain interlocks;
- Bypassed existing LLRF chain, RF Safety Crate output connected directly to preamplifier (222S).

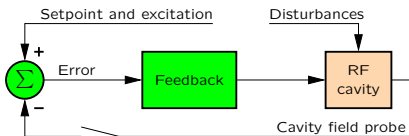


Progress

- Monday
 - Connected LLRF9 to inputs only, established signal levels and transferred calibrations;
 - Connected drive output, configured feedback loops;
 - Established closed-loop operation in CW mode.
- Tuesday
 - Interfaced LLRF9 tuner control loops to booster motor control;
 - Established closed-loop operation of tuner and field balance loops.
- Wednesday
 - Transitioned to closed-loop operation with ramping and beam;
 - Characterization of open and closed-loop spectra.
- Thursday
 - Step response measurements;
 - System set up from scratch by Alun Watkins.
- Friday
 - Performance testing;
 - Discussion.



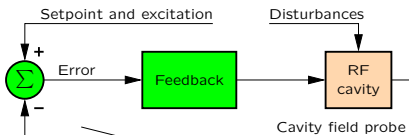
Open Loop Transfer Function



- Measured from setpoint to the cavity probe;
- Feedback block in open loop has no dynamics, just gain and phase shift;
- Open loop cavity response;
- Fit resonator model to extract gain, loaded Q , detuning, delay, phase shift at ω_{rf} ;
- Faster than expected gain roll-off above the resonance.



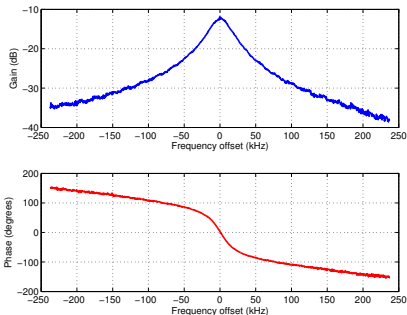
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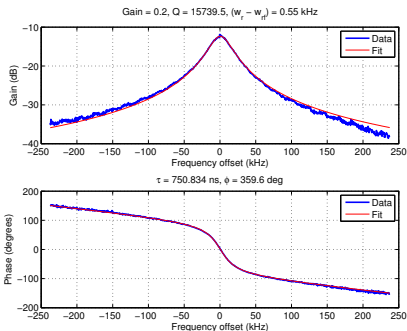
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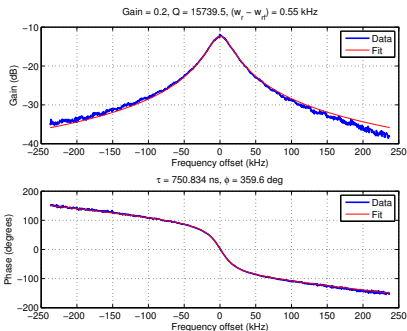
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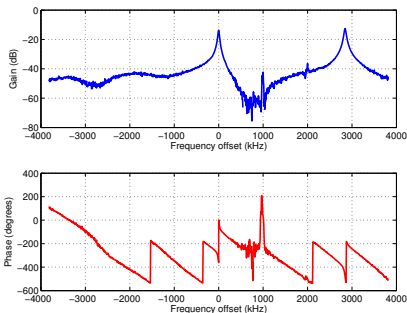
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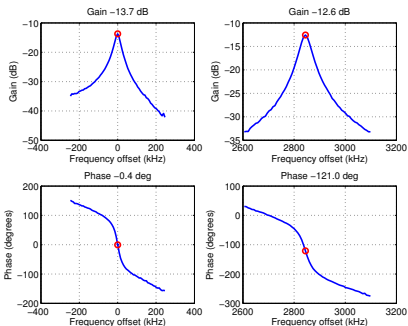
Wideband Open Loop Transfer Function



- Wider sweep reveals a parasitic mode at 2.8 MHz above the π mode;
- Negative feedback for the π mode is positive for the parasitic mode;
- This positive feedback limits direct loop gain;
- The simplest way around the issue is to use digital delay to equalize the modal phase shifts (230 ns).



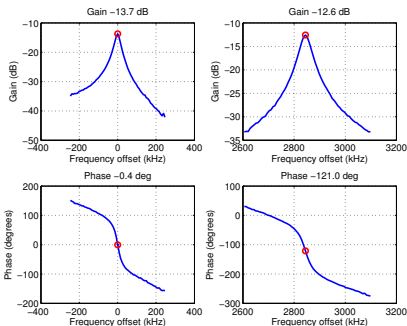
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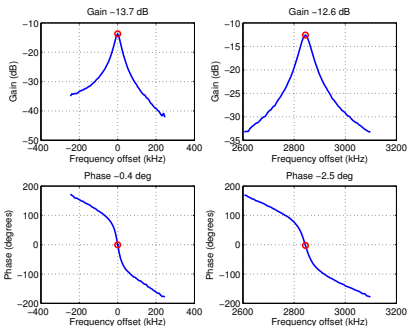
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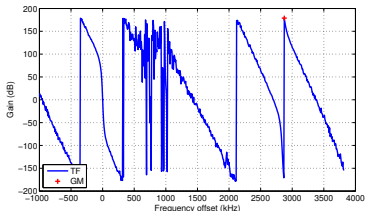
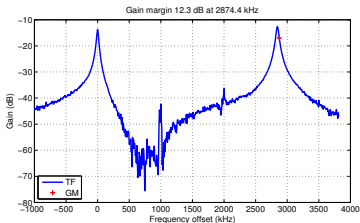
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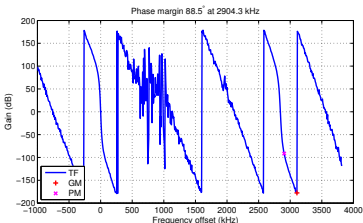
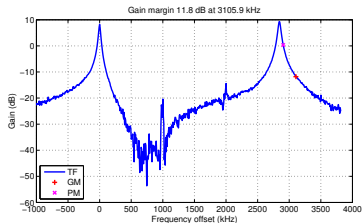
Proportional Loop Gain and Delay



- Set up minimum delay and equalized transfer functions for identical 3 dB closed-loop peaking.
 - Minimum delay: peak gain at RF is -9.2 dB, gain margin 12.3 dB
 - Equalized: peak gain at RF is $+8$ dB, gain margin 11.8 dB, phase margin 88 degrees
- More sophisticated parasitic mode suppression methods can improve the performance only slightly, around 2-3 dB.



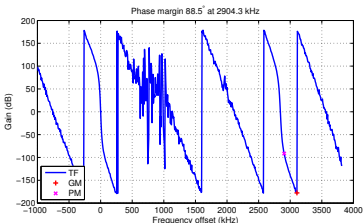
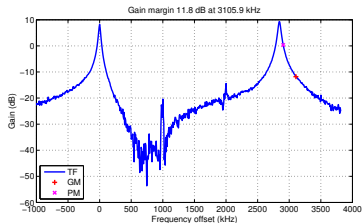
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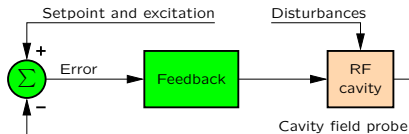
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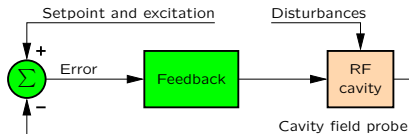
Closed Loop Transfer Function



- Measured from setpoint to the error signal;
- Quantifies closed-loop disturbance rejection vs. frequency offset from f_{RF} ;
- Proportional and integrator loops produce high rejection at low frequencies;
- Magnitude on log-log scale, field setpoint of 1 MV.



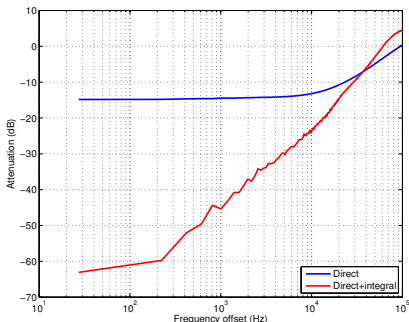
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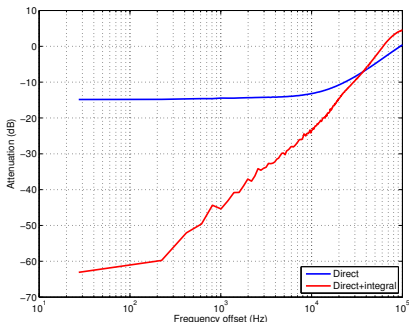
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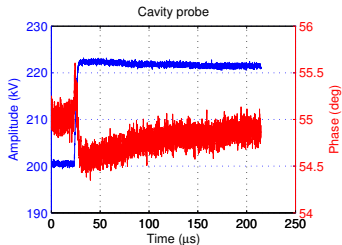
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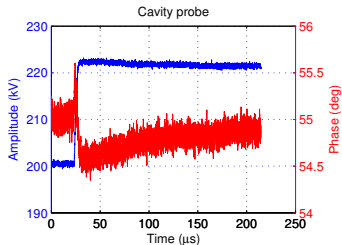
Step Response



- Use ramp start signal to trigger waveform acquisition;
- Ramp profile loaded with a 10% amplitude step (200 to 220 kV);
- Some coupling between amplitude and phase;
- All input channels are captured on the same trigger, observing reflected power here;
- Characterized a 5 degree phase step as well.



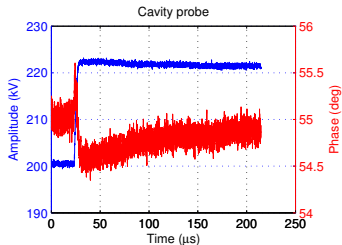
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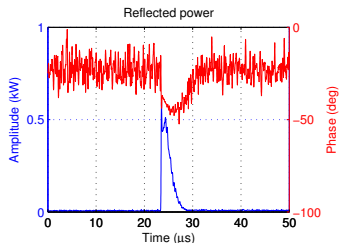
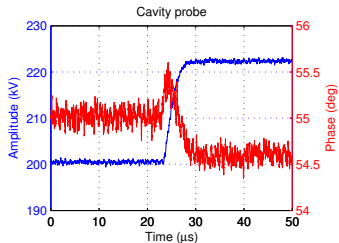
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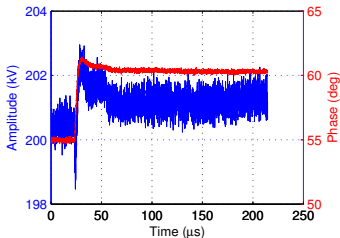
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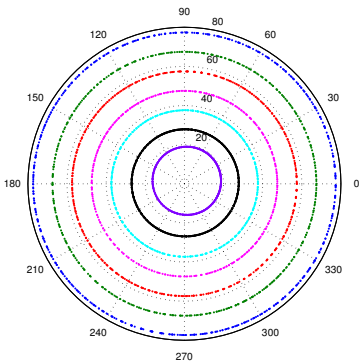


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Frequency offset input, IQ circles, decibel radius scale

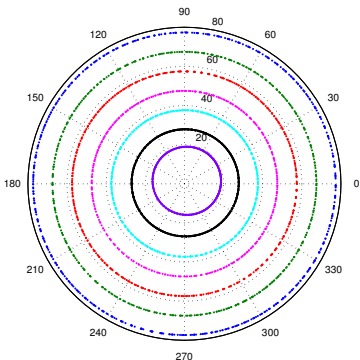


- Input signal at -60 to 0 dBm in 10 dB steps, ≈ 1 Hz offset;
- Log scale polar plot ($r = 20 \log_{10}|x|$);
- Center offset, more prominent at low amplitudes, is due to RF feedthrough (coupled from the reference channel);
- Can numerically remove the offsets;
- Good linearity (uncorrected for synthesizer errors), consistent offsets at -73 dBFS.



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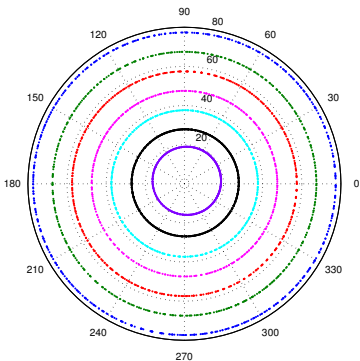


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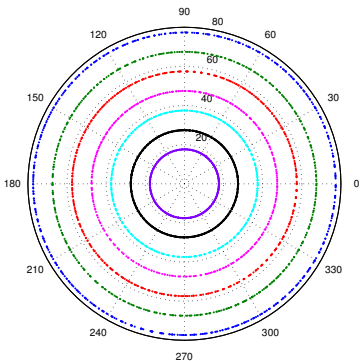


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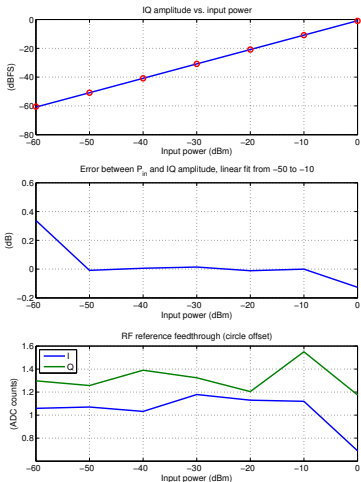
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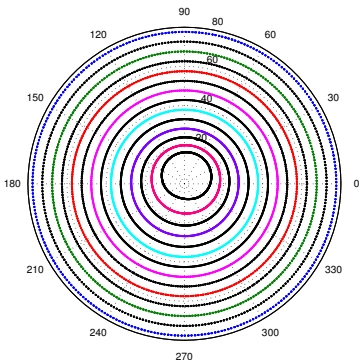


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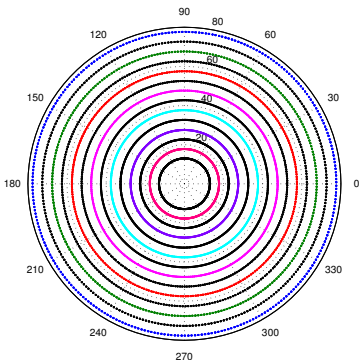


- Similar to the test at DLS with several modifications:
 - Reference level at +10.5 dBm vs. +8 dBm at DLS;
 - Input power level monitored by a spectrum analyzer;
 - 5 dB steps.
- Center offset correction;
- Linearity good to ± 0.05 dB down to -50 dBFS;
- Offsets/feedthrough at -72 dBFS.



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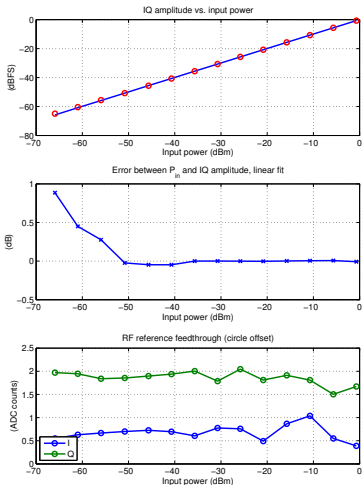
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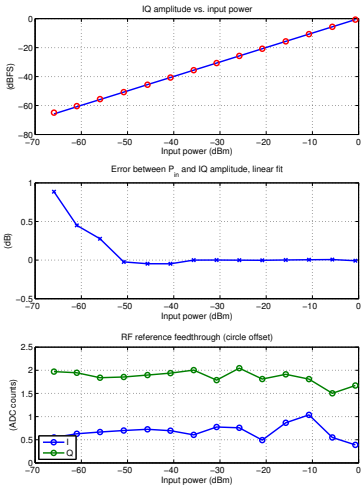
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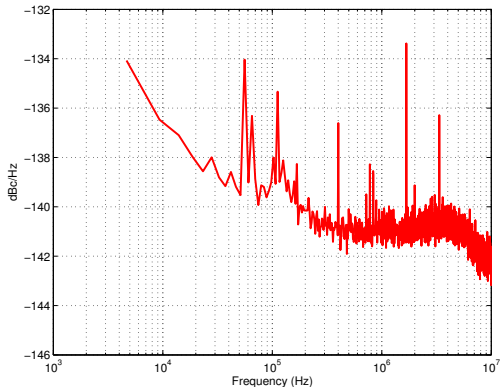
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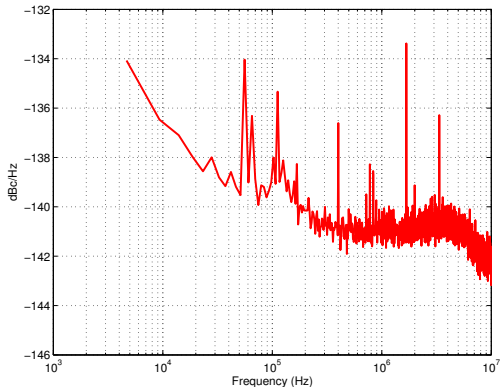
Single Channel Phase Noise



- Reference channel at DLS, +8 dBm input level;
- Sources of lines at 65.2 and 111.8 kHz are known, already fixed;
- Same channel measured on the bench, HP8664A at +11.5 dBm.



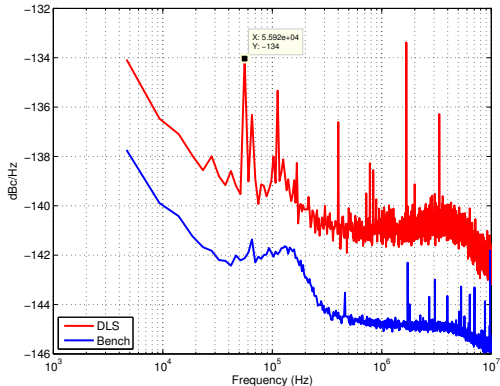
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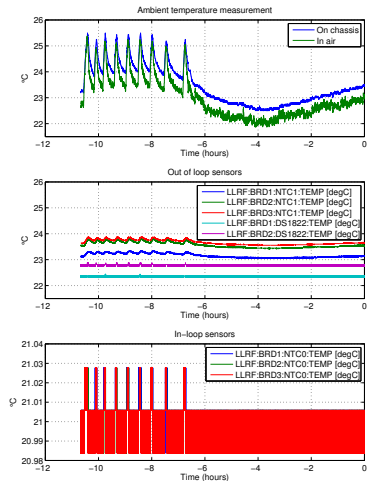
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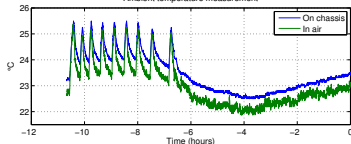


- 9 internal sensors on cold plate: 6 NTCs, 3 DS18B20 digital sensors;
- Three temperature stabilization loops using thermoelectric coolers;
- Two external sensors, in air and attached to chassis;
- Tight stabilization of in-loop sensors;
- Residual sensitivity of out-of-loop sensors is $0.09\text{--}0.12\text{ }^{\circ}\text{C}/^{\circ}\text{C}$.

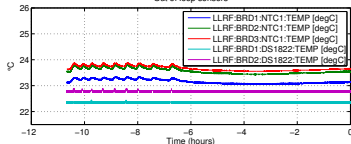


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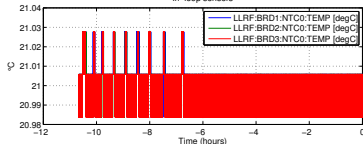
Ambient temperature measurement



Out of loop sensors



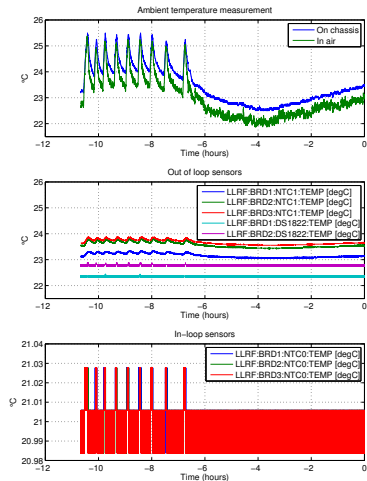
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- Tight stabilization of in-loop sensors;
- Residual sensitivity of out-of-loop sensors is 0.09–0.12 °C/°C.



Thermal Stability

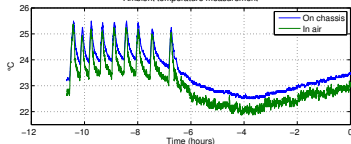


- 9 internal sensors on cold plate: 6 NTCs, 3 DS18B20 digital sensors;
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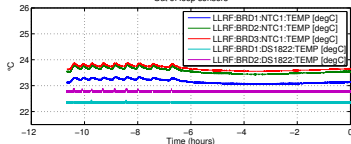


Thermal Stability

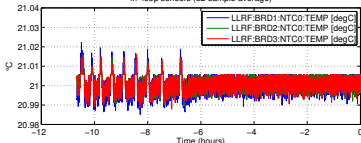
Ambient temperature measurement



Out of loop sensors



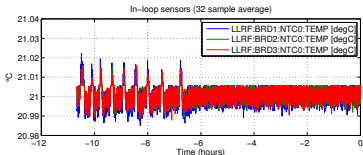
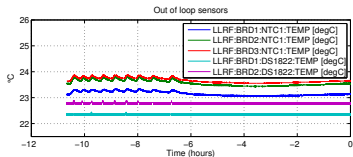
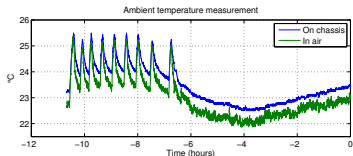
In-loop sensors (32 sample average)



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Summary

- **Successfully operated booster RF station with beam;**
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- Demonstrated field control at forward power 60 W to 60 kW;
- Showed high rejection of disturbances, good tuner loop response with low reflected power.



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